

Model 134A27

Tourmaline ICP® pressure bar, 20k psi, 0.25 mV/psi, 0.2 microsecond rise

# Installation and Operating Manual

This manual contains the 402A17 installation and operating manuals that comprise a Model 134A27 Tourmaline ICP® pressure bar, 20k psi, 0.25 mV/psi, 0.2 microsecond rise time kit.

For assistance with the operation of this product, contact PCB Piezotronics, Inc.

Toll-free: 800-828-8840 24-hour SensorLine: 716-684-0001 Fax: 716-684-0987 E-mail: info@pcb.com Web: www.pcb.com







# The information contained in this document supersedes all similar information that may be found elsewhere in this manual.

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**Service** – Due to the sophisticated nature of the sensors and associated instrumentation provided by PCB Piezotronics, user servicing or repair is not recommended and, if attempted, may void the factory warranty. Routine maintenance, such as the cleaning of electrical connectors, housings, and mounting surfaces with solutions and techniques that will not harm the physical material of construction, is acceptable. Caution should be observed to insure that liquids are not permitted to migrate into devices that are not hermetically sealed. Such devices should only be wiped with a dampened cloth and never submerged or have liquids poured upon them.

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**Calibration** – Routine calibration of sensors and associated instrumentation is

recommended as this helps build confidence in measurement accuracy and acquired data. Equipment calibration cycles are typically established by the users own quality regimen. When in doubt about a calibration cycle, a good "rule of thumb" is to recalibrate on an annual basis. It is also good practice to recalibrate after exposure to any severe temperature extreme, shock, load, or other environmental influence, or prior to any critical test.

PCB Piezotronics maintains an ISO-9001 certified metrology laboratory and offers calibration services, which are accredited by A2LA to ISO/IEC 17025, with full traceablility to N.I.S.T. In addition to the normally supplied calibration, special testing is also available, such as: sensitivity at elevated cryogenic temperatures, phase or extended response, high or low frequency response, extended range, leak testing, hydrostatic pressure testing, and others. For information on standard recalibration services or special testing, contact your local PCB Piezotronics distributor, sales representative, or factory customer service representative.

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DOCUMENT NUMBER: 21354 DOCUMENT REVISION: B ECN: 17900

#### SERIES 134A20 TOURMALINE PRESSURE BAR GENERAL OPERATION MANUAL

### **1.0 INTRODUCTION**

Series 134A20 Tourmaline Pressure Bar with In-Line Amplifier consists of a Series 134A Tourmaline Pressure Bar connected to a three-inch, low-noise cable that connects into Series 402 In-Line Source Follower Voltage Amplifier. See Figure 1.



The Tourmaline Pressure Bars is a high-pressure, fast rise time, suppressed-resonance blast sensor ideal for single-shot, high-frequency measurements of incident or reflected shock wave pressures found in studies of plasma physics and hyper-sonics. Ranges up to 20,000 psi are available.

The Series 134A20 is typically powered by a PCB Series 480 ICP<sup>®</sup> Power Conditioner (constant current) or equivalent. Power and signal is conducted over a single-conductor coaxial cable with the shield serving as a signal return. Output from the power conditioner is usually fed into a high-speed readout instrument, such as a digital oscilloscope.

### 2.0 DESCRIPTION

The Series 134A20 Tourmaline Pressure Bar with In-Line Amplifier consists of three components attached together and calibrated as a system. The Series 134A Tourmaline Pressure Bar senses and transduces a pressure/shock wave into a highimpedance, electrostatic charge. This charge is sent through the three-inch, low-noise cable into the Series 402A In-Line Amplifier, where it is converted into a low-impedance voltage with less than 100 ohms output impedance. See Figure 2. The circuit in the in-line amplifier consists of a MOSFET input IC with a very high transconductance, resulting in a voltage gain that is very close to unity. The source terminal "follows" the gate voltage instantaneously, without a shift in phase (i.e., the amplifier is non-inverting).



## 2.1 SYSTEM OUTPUT

The system output of Series 134A20 is a lowimpedance output governed by the following equation:

$$V = \frac{Q}{C1 + C2 + C3}$$

where:

- V = voltage sensitivity of system (V/psi)
- Q = charge output of Tourmaline Bar (pC/psi), Model 134A
- C1 = capacitance of Tourmaline Bar (pF), Series 134A
- C2 = capacitance of three-inch cable (pF), Model 003A000-03
- C3 = capacitance of input of in-line amplifier (pF), Series 402A

A system calibration is supplied with Series 134A20 that shows the system output in mV/psi.

## 3.0 INSTALLATION

See the Installation Drawing included in the back of this manual for a description of the Tourmaline Bar and its installation. *Install in a gas medium where the measurement is to be taken. The sensor can not be normally used in a liquid medium.*  1

Connect the Series 134A20 Tourmaline Pressure Bar with In-Line Amplifier to a PCB power conditioner. Connect the output of the power conditioner to an appropriate readout instrument.



Tie the three-inch, low-noise cable to a rigid structure to prevent excessive motion and noise. Allow for strain relief.

### Caution

If the tourmaline pressure bar or the three-inch cable become disconnected from the in-line amplifier, take caution to momentarily electrically short out the cable and tourmaline pressure bar before reconnecting the cable to the in-line amplifier. This prevents any unusual high charge buildup from damaging the input stage of the inline amplifier.

Black vinyl tape is supplied on the sensing end of the tourmaline bar to dampen heat transfer into the sensing crystal. This tape can be removed and replaced if damaged.

## 4.0 **OPERATION**

Switch on the (PCB) power conditioner and observe the bias monitoring voltmeter (or LED) on the front panel.

If the indicator reads proper bias (green area on the meter, approximately 11 volts), the connections from the Series 402A In-Line Amplifier are correct and the in-line amplifier is operational. If the meter reads in the red or yellow areas, the system is shorted or open. The short or open circuit may be in the in-line amplifier, the cable connecting the in-line amplifier to the power conditioner, or in the power conditioner.

Allow the system to stabilize for a minute. A signal drift may occur when connecting the power conditioner to the readout instrument. This should stabilize as components in the power conditioner and readout instrument charge.

If long cables between the Series 134A20 and the power conditioner are used, impedance matching of the long cable and increase of the constant current in the power conditioner may be required because of the high-speed response of the sensor. Check with PCB if there are questions.

# 5.0 CALIBRATION

The tourmaline pressure bar must be calibrated *dynamically*. The tourmaline pressure bar does not measure static pressures. In addition, application of high static pressure to the tourmaline bar may cause damage.

Two methods are used at PCB to calibrate the Series 134A20: a shock tube and a hydraulic pulse tester. The shock tube is used primarily to check the sub-microsecond response of the tournaline pressure bar.

The hydraulic pulse tester is used to generate a known, short-duration pressure pulse at five different pressure levels to determine the average sensitivity of the sensor.

Recalibration services are offered by PCB. In addition, the shock tube and hydraulic pulse calibrator are offered as standard products by PCB.

# 6.0 USE OF CHARGE MODE TOURMALINE PRESSURE BAR

The tourmaline pressure bar is a charge output piezoelectric sensor that can be used without the inline source follower amplifier for certain applications. Calibration information is supplied showing (charge) output sensitivity of the tourmaline bar alone, along with its crystal capacitance.

#### SERIES 134A20 TOURMALINE PRESSURE BAR GENERAL OPERATION MANUAL

Note that the Series 402A In-Line Amplifier features a frequency response of greater than 1 MHz. A charge amplifier used with the tourmaline pressure bar may limit the upper frequency response of the system.

Connecting the Series 134A Tourmaline Pressure Bar directly to a readout instrument without the use of the Series 402A In-Line Amplifier or other impedance-converting device can be done with some limitations. The input impedance of the readout instrument may be significantly low to create a very high-pass filtering effect. This can affect measurement of pulse duration or amplitude of a slow-rising pulse. Also, the voltage sensitivity of the output of the tourmaline bar into the readout instrument is affected by the cable capacitance and input capacitance of the readout instrument.

Contact the factory with any questions regarding use of the sensor in the charge output mode.

## 6.1 USE OF THE IN-LINE AMPLIFIER

The Series 402A In-Line Amplifier is used to convert the high-impedance output of the tourmaline pressure bar into a low-impedance output. The in-line amplifier may also be used as a unity gain amplifier for checking system frequency response. A signal generator is connected to the input of the in-line amplifier and fed through the system.

# 7.0 MAINTENANCE AND REPAIR

The electrical connectors on the tourmaline pressure bar, the input of the in-line amplifier, and the three-inch, low-noise cable must be kept clean and dry, especially if they are operating in a dusty or wet environment. This is to prevent drift due to low insulation resistance. Also, note the following precautions:

1. Do not exceed maximum levels.

2. Constant temperatures around the tourmaline bar should be limited to 150°F. Short-term exposure to higher temperatures, such as that found in a blast wave, do not cause damage.

3. Use only PCB power conditioners or approved alternative.

4. Current to the in-line amplifier must be limited to not more than 20 mA to avoid damage.

5. If the tourmaline pressure bar and in-line amplifier are to be left outside overnight or in humid or rainy environments, they should be coated with silicone oil and covered.

				IDE	SENGOD	Revision: G
					SENSUR	ECN #: 29109
Performance	E	NGLISH	SI		OPTIONAL VERSIONS	
Measurement Range(for ±5V output)		20 kpsi	137,900 kPa		Optional versions have identical specifications and accessories as lister	d for the standard model
Sensitivity(± 15 %)		.25 mV/psi	0.04 mV/kPa	[1]	except where noted below. More than one option may be	be used.
Maximum Pressure(static)		20 kpsi	137,900 kPa			
Resolution		300 mpsi	2.1 kPa	[2]	M - Metric Mount	
Resonant Frequency		1500 kHz	≥ 1500 kHz			
Rise Time(Reflected)		0.2 µ sec	≤ 0.2 µ sec		N - Negative Output Polarity	
Low Frequency Response(-5 %)		1 Hz	1 Hz			
Non-Linearity	:	≤ 2 % FS	≤ 2 % FS	[3]		
Environmental						
Temperature Range(Operating)	+32	2 to +120 °F	0 to +49 °C			
Maximum Flash Temperature		5000 °F	2760 °C		NOTES:	ter dan menahan darim 1999 nang dan dan dari bertakan dari bahar dari bahar dari bahar dari bahar dari bahar da
Electrical					[1] Calibrated as a system using (3 inch) type 003 cable and 402A serie	s in-line amplifier.
Output Polarity(Positive Pressure)		Positive	Positive		[2] Typical.	
Discharge Time Constant(at room temp)		≥ 0.5 sec	≥ 0.5 sec		[3] Zero-based, least-squares, straight line method.	
Excitation Voltage		to 30 VDC	20 to 30 VDC		[4] See PCB Declaration of Conformance PS023 for details.	
Constant Current Excitation		to 20 mA	2 to 20 mA			
Output Impedance	≤	100 ohm	≤ 100 ohm			
Output Bias Voltage		to 14 VDC	8 to 14 VDC			
Physical						internet of the control theory of the second dependence of
Sensing Element		ourmaline	Tourmaline			
Housing Material		inless Steel	Stainless Steel			
Diaphragm		Epoxy	Epoxy		SUPPLIED ACCESSORIES	
Sealing		Epoxy	Epoxy		Model 061A30 Spanner Wrench, 2 Pin (1)	
Electrical Connector	10-32	Coaxial Jack	10-32 Coaxial Jack			
Weight		1.4 oz	39 gm			
					Entered: BS Engineer:RF Sales: RWM Approved:R	M Spec Number:
(6					Date: 1608 Date: 1608 Date: 1608 Date: 1608	6617
All specifications are at room temperature unless otherwise specified. In the interest of constant product improvement, we reserve the right to change specifications without notice.					<b>PCB</b> PIEZOTRONICS Phone: 7' Fax: 716-	16-684-0001 686-9129
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Model 402A17

**In-line Amplifier** 

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DOCUMENT NUMBER: 21354 DOCUMENT REVISION: B ECN: 17900

### OPERATING GUIDE

SERIES 401A & 402A

### ICP SOURCE FOLLOWER AMPLIFIERS



The series 401Aand 402A are miniature, unity gain, impedance converting voltage amplifiers which operate from constant current power units in a 2-wire mode.

These epoxy potted amplifiers are designed to convert the high impedance voltage from piezoelectric transducers to a low impedance voltage able to drive most readout instruments directly.

The input impedance of these amplifiers is extremely high  $(10^9 \& 10^{11} \text{ ohms})$  while the output impedance is less than 100 ohms.

Except for mechanical configuration, the 401A series is identical to the 402A series.

See Guide G-0001 for a complete coverage of the ICP (Integrated Circuit Piezoelectric) concept.

### 2.0 DESCRIPTION

The series 401A is the "connector" series, i.e. it is designed to attach directly to the TO-32 co-axial connector of piezoelectric transducers as shown in Figure 1.



Figure 1 Typical Application Series 401A



CONNECTOR TYPE 401A Voltage-Follower IN-LINE TYPE 402A Voltage-Follower

When used as in Figure 1, the piezoelectric transducer is converted to ICP operation.

The series 402A "in-line" series is designed to be used in a co-axial line as shown in Figure 2.



SCHEMATIC DIAGRAM

Figure 2 Typical Application Series 402A

The series 402A is useful in field applications where high temperature or other severe environmental factors precludes the use of transducers with built-in amplifiers.

# 2.1 THEORY OF OPERATION

These impedance converting amplifiers utilize a MOSFET input integrated circuit as shown in Figures 1 & 2.

The circuit consists of a MOSFET input IC with a very high transconductance (gm) resulting in a voltage gain of very close to unity. The source terminal "follows" the gate voltage instantaneously and without a shift in phase, i.e. the amplifiers are non-inverting.

The input resistor, R serves 2 functions.

It biases the gate at drain potential for proper operation of the IC device and : it drains off spurious long term charge that exists at the input terminal from thermal effects on crystal transducer elements.

Because of the lowered insulation resistance of ceramic and non-quartz crystal transducers, this input resistance should be  $10^9$ ohms when used with such transducers. Consult our data sheets for the appropriate model numbers. (See section 7.0)

Quartz transducers on the other hand, have very high internal resistance and low capacitance, thus, the 1011 value of input resistance should be used for this type of transducer.

In both models, the outer case is at electrical ground potential.

# 2.2 INPUT DISCHARGE TIME CONSTANT

The input time constant is dependent upon two parameters, 1) the input resistor value 2) the total input capacitance.

The input capacitance value is the sum of all capacitance across the input connector, including transducer capacitance, cable capacitance (if used) and input ranging capacitor (C).

The input resistance is either 10<sup>9</sup> ohm or 10<sup>11</sup> ohms depending upon exact model ordered.

To calculate the resultant discharge time constant (Sec), simply multiply the input resistor value (ohms) by the total shunt capacitance (farads).

Refer to Guide GOOO1 for effect of discharge TC on low frequency response.

### 2.3 RANGING

The voltage sensitivity of the transduction system varies inversely with the total input shunt capacitance and directly with the transducer charge sensitivity as follows:

(Eq. 1)

V = Q $C_t$ 

where:

Q = charge sensitivity of transducer in pC/mech. unit.

 $C_t$  = Total shunt capacitance in pF. \*

V = Voltage sensitivity in volts/mech. unit. To obtain rough value of system sensitivity, simply divide the transducer charge sensitivity by total capacitance at the amplifier input as shown in Eq. 1.

Note: Don't forget to include input cable capacitance when using the series 402A.

To increase the range of a transducer i.e. to decrease system voltage sensitivity, select a 401A or 402A with larger value of input capacitor. 3 values are available; 10, 100 and 1000 pf.

Final system sensitivity should always be determined by calibration with known inputs.

### 3.0 INSTALLATION

The Model 401A is simply threaded directly on to the 10-32 coaxial connector of most transducers.

To exclude moisture from this high impedance input connection, protect joint with shrink tubing, RTV rubber coating or other suitable means.

Connect the power/signal connector to an ICP Power Unit such as the Model 480B, 482A, 484B, 483A etc. Using co-axial cable or solder connector adaptors (Model 070A09) and ordinary 2-wire cable.

In vibratory environments, clamp 402A to surface to prevent damage due to excessive shock motion. A plastic clamp is provided for this purpose.

Connect the power/signal connector to power unit as previously described for Model 401A.

Caution: Before connecting inputs to both models, short out cable and/or transducer to dissipate any accumulated charges.

# 4:0 OPERATION

Operation of these amplifiers is very similar to operation of ICP transducers.

All PCB Power Units have fault monitor meters on the front panel to monitor operation of the built-in amplifiers in ICP transducers. Since the Models 401A and 402A utilize the identical circuit as ICP transducers, use of the fault monitor meter is similar. A mid-scale reading on the meter (green area) indicates proper bias level for the amplifier.

A zero reading (red area) indicates a shorted condition in the built-in amplifier or in the cable connecting amplifier to power unit.

A full scale reading (yellow area) indicates an open circuit condition in amplifier or interconnecting cable.

### 5.0 FREQUENCY RESPONSE

The low frequency response of these amplifiers when used with piezoelectric transducers is dependent upon the discharge time constant (or coupling time constant of power unit, if shorter than the discharge TC) in accordance with the following relationship:

 $f_0 = \frac{.16}{RC}$ 

(Eq 2)

where:

fo = -3db low freq. cut-off, (Hz)

RC = input discharge TC

See section 6.1.1 of Guide G-0001 for a more thorough discussion of this topic.

The high frequency response of these amplifiers, for small signals and driving short cables is better than 1 megahertz.

For optimum frequency response when driving long cables at maximum output voltages, use maximum allowable input current (20mA).

Many PCB Power Units such as 482A & 484B have adjustable output current to facilitate driving long lines at high voltage levels.

In most cases, high frequency response in actual use will be limited by transducer parameters.

### 6.0 MAINTENANCE

The miniature size and epoxy encapsulation precludes most maintenance. Should input connectors become dirty, reducing input time constant, clean with lint free wipes dipped in freon TF (R) or other suitable solvent. Avoid solvent containing ketones.

Caution: Avoid touching the input terminal directly with the fingers since accumulated static charge can destroy the input MOSFET.

After wiping input connector, dry in oven at  $200^{\circ}$ F for several hours. Do not heat beyond  $+250^{\circ}$ F.

### 7.0 CAUTION - HIGH OUTPUT, THERMALLY SENSITIVE TRANSDUCERS

Certain ceramic transducers are capable of generating huge quantities of charge (and voltage) when affected by relatively small thermal transients such as when picked up by hand.

In recent tests at PCB, a ceramic accelerc meter, when picked up and held in the hand produced over 100,000 pC of charge resulting in over 100 volts into a high impedance load, over a period of several seconds.

High voltages like this can destroy the input mosfet in the 401A and 402A if not drained away fast enough.

For transducers such as these, it may be necessary to reduce the input resistor below 1 x 10<sup>9</sup> ohms (1 x 10<sup>8</sup> or 1 x 10<sup>7</sup>) to rapidly dissipate such spurious outputs. Generally, internal capacitances of these types of transducers are extremely high (> 1000 pF) resulting in adequate low frequency response, even with the low value input resistance. Consult factory if it is decided this problem exists in your installation.

### MANUAL NUMBER: 18429 MANUAL REVISION: NR

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Model Number				IEIED	Revision: D		
402A17			IN-LINE AN	ECN #: 29854			
Performance		ENGLISH	SI		OPTIONAL VERSIONS		
Input Range		± 5.0 V	+ 5.0 V		Ontional versions have identical specifications and accessories as liste	listed for the standard model	
Overrange		± 10 V	± 10 V	[2]	except where noted below. More than one option may	av be used.	
Output Range(AC)		± 5.0 V	± 5.0 V				
Voltage Gain(± 2 %)(non-inverting)		0.98	0.98				
Low Frequency Cutoff(-5 %)		1.0 Hz	1.0 Hz	[3]			
High Frequency Response		1000 kHz	1000 kHz	[4]			
Environmental							
Temperature Range(Op	Temperature Range(Operating)		-54 to +121 °C				
Maximum Shock		5000 g pk	49,050 m/s <sup>2</sup> pk				
Maximum Vibration(5 to 2000 Hz)		1000 g pk	9810 m/s² pk				
Electrical							
Excitation Voltage	Excitation Voltage		18 to 28 VDC		NOTES:		
Constant Current Excitat	tion	2 to 20 mA	2 to 20 mA		[1] Actual Discharge Time Constant is the product between the Input Resistance and the Total		
Output Bias Voltage		8 to 14 VDC	8 to 14 VDC 8 to 14 VDC		Input Capacitance (including cable capacitance and sensor capacita	ance).	
Input Resistance(± 20 %	.)	10 <sup>9</sup> ohm	10 <sup>9</sup> ohm		[2] For +10 volt output, minimum 24 VDC supply voltage required. Neg	ative 10 volt output may	
Input Capacitance(± 10 %)		500 pF	500 pF		131 Actual Low Frequency Cutoff is approximately 0.5 + Actual Dischart	ge Time Constant	
Maximum Input Voltage(without damage)		100 V	100 V		[4] High frequency response may be limited by supply current and out	put cable length.	
Discharge Time Constant(± 30 %)		0.5 sec	0.5 sec	[1]	[5] See PCB Declaration of Conformance PS024 for details.		
Output Impedance		<100 ohm	<100 ohm				
Broadband Electrical Noise(1 to 10 kHz)		43.0 µV	43.0 µV				
Physical							
Housing Material		Stainless Steel	Stainless Steel				
Sealing		Welded Hermetic	Welded Hermetic				
Electrical Connector(Inp	ut)	10-32 Coaxial Jack	10-32 Coaxial Jack				
Electrical Connector(Out	iput)	10-32 Coaxial Jack	10-32 Coaxial Jack				
Size (Diameter x Length	Size (Diameter x Length)		6.4 mm x 29.7 mm				
Weight		0.132 oz	3.75 gm			~~	
					Entered S Engineer S A Sales: PWN Approved	Spec Number:	
					Date 8 8 Date: 2 6 8 Date: 2 10 Date: 2 10	402-1170-80	
					44		
151 (5)							
All energifications are at -	com tomporatura unloss athensi	ion apposified		/16-684-0001			
In the interest of constant	t product improvement, we rese	rve the right to change speci	<b>FIGHE / 10-004-001</b>				
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